

The Relationship Between Flux Density and Brightness Temperature

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Contents

1	The Answer	1
2	Introduction	1
3	Elliptical Gaussian Source	3
4	Uniform Disk Source	5

1 The Answer

In this document I derive the general relationship between the flux density of a source and its image brightness temperature for an antenna beam measuring two different source geometries, uniform disk and gaussian. For the impatient I give the answer here. For those interested in the details, subsequent sections provide those details. For a point source, measured with a gaussian antenna beam from an antenna with main reflector diameter D , the relationship between the source's flux density ($S_\nu^{point}(Jy)$) and its Rayleigh-Jeans equivalent brightness temperature (T_A) is given by:

$$\begin{aligned} S_\nu^{point}(Jy) &= \frac{\pi k}{2(\ln(2)D)^2} \frac{T_A}{\eta_{mb}} \\ &\simeq 4513.90 \frac{T_A}{D^2(m)\eta_{mb}} (K) \end{aligned} \tag{1}$$

2 Introduction

In the following I derive the relationship between the flux density of a source and its brightness temperature for an antenna beam measuring two different source geometries,

uniform disk and gaussian. The general relation between the flux density of a source and its brightness temperature is

$$S_\nu = \frac{2k}{\lambda^2} \int T_B(\Omega) d\Omega \quad (2)$$

Note that Equation 2 assumes that the Rayleigh-Jeans approximation ($h\nu \ll kT$) applies. In a way, then, this is a fictitious temperature which is useful for describing the output power from a radio source using a relation that is linearly proportional to temperature. For a source with brightness temperature distribution $\psi(\xi, \eta)$, we can define the source solid angle as follows

$$\Omega_s = \int \int \psi(\xi, \eta) d\xi d\eta. \quad (3)$$

The solid angle of the source normalized by the primary beam pattern of the antenna ($f(\xi, \eta)$), sometimes called the “effective source solid angle”, is given by

$$\Omega'_s = \int \int f(\xi, \eta) \psi(\xi, \eta) d\xi d\eta, \quad (4)$$

while the main beam solid angle is given by

$$\Omega_m = \int \int_{mb} f(\xi, \eta) d\xi d\eta, \quad (5)$$

and the full beam solid angle is given by

$$\Omega_A = \int \int_b f(\xi, \eta) d\xi d\eta. \quad (6)$$

As the main beam full-width at half-maximum (FWHM) beam width θ_B is defined as follows (see Baars 2007, “The Paraboloidal Reflector Antenna in Radio Astronomy and Communication”, Chapter 4):

$$\theta_B = \frac{b\lambda}{D}, \quad (7)$$

it is convenient in the following to parameterize the main beam solid angle in terms of the illumination taper employed by the measurements:

$$\begin{aligned} \Omega_m &= \frac{\pi}{4} \left(\frac{b\lambda}{D} \right)^2 \\ &= \frac{\pi b^2 \theta_{Ba} \theta_{Bb}}{4}, \end{aligned} \quad (8)$$

where b is the illumination taper factor. For a Gaussian beam $b = 1/\sqrt{\ln(2)} \simeq 1.2$.

With measurements of the solid angle of the source normalized by the primary beam pattern of the antenna (Ω'_s), the primary (main) beam (Ω_m), and the main beam efficiency (η_{mb}), we can write the relation between the measured antenna temperature (T_A) and the source brightness temperature (T_B) as follows

$$T_A = \frac{\Omega'_s}{\Omega_m} T_B \eta_{mb} \quad (9)$$

Substituting for the quantity $\int T_B d\Omega_s = T_B \Omega_s$ in Equation 2 results in the following general relation between flux density and measured antenna temperature

$$\begin{aligned} S_\nu &= \frac{2k}{\lambda^2} T_A \Omega_A \frac{\Omega_s}{\Omega'_s} \\ &= \frac{2k}{\lambda^2} \frac{T_A}{\eta_{mb}} \Omega_m \frac{\Omega_s}{\Omega'_s} \end{aligned} \quad (10)$$

where we have defined the main beam efficiency η_{mb} as follows:

$$\eta_{mb} \equiv \frac{\Omega_m}{\Omega_A} \quad (11)$$

Note that the term $\frac{\Omega_s}{\Omega'_s}$ defines the coupling of the measured source to the beam of the antenna.

In the following I calculate $\frac{\Omega_s}{\Omega'_s}$ for two standard source distributions. For these calculations I use the simplification that the antenna beam is circular ($\theta_{Ba} = \theta_{Bb} = \theta_B$). This is a common situation in practice, and avoids the need to include a position angle term in our beam and source major (θ_{Ba} and θ_{Sa}) and minor (θ_{Bb} and θ_{Sb}) axis terms¹.

3 Elliptical Gaussian Source

For an elliptical Gaussian source with dimensions θ_{Sa} θ_{Sb} and an elliptical beam with FWHM θ_B ,

$$\begin{aligned} \Omega_S &= \int \int_{-\infty}^{\infty} \exp \left[-4 \ln(2) \left(\frac{\xi^2}{\theta_{Sa}^2} + \frac{\eta^2}{\theta_{Sb}^2} \right) \right] d\xi d\eta \\ &= \frac{\pi \theta_{Sa} \theta_{Sb}}{4 \ln(2)} \end{aligned} \quad (12)$$

and

$$\begin{aligned} \Omega_m &= \int \int_{-\infty}^{\infty} \exp \left[-\frac{4}{b^2} \left(\frac{\xi^2}{\theta_{Ba}^2} + \frac{\eta^2}{\theta_{Bb}^2} \right) \right] \\ &= \frac{\pi b^2 \theta_B^2}{4}. \end{aligned} \quad (13)$$

With Equations 12 and 13 the solid angle of the source multiplied by the primary beam pattern of the antenna becomes

$$\begin{aligned} \Omega'_S &= \int \int_{-\infty}^{\infty} \exp \left[-4 \ln(2) \left(\frac{\xi^2}{\theta_{Sa}^2} + \frac{\eta^2}{\theta_{Sb}^2} \right) \right] \exp \left[-\frac{4}{b^2} \left(\frac{\xi^2 + \eta^2}{\theta_B^2} \right) \right] d\xi d\eta \\ &= \frac{\pi b \theta_B^2}{4 \sqrt{\ln(2)}} \left(\frac{\theta_{Sa}^2}{\theta_{Sa}^2 + \theta_B^2} \right)^{1/2} \left(\frac{\theta_{Sb}^2}{\theta_{Sb}^2 + \theta_B^2} \right)^{1/2}, \end{aligned} \quad (14)$$

¹This entails including the scaling factor $\cos PA$ in all equations that include major and minor axis variables.

where I have used the fact that

$$\int_0^\infty \exp(-r^2 x^2) dx = \frac{\sqrt{\pi}}{2r}. \quad (15)$$

We can now write the source correction factor defining the coupling between a Gaussian source and a Gaussian beam as follows

$$\begin{aligned} \frac{\Omega_s}{\Omega'_s} &= \frac{\frac{\pi}{4 \ln(2)} (\theta_{Sa} \theta_{Sb})}{\frac{\pi b \theta_B^2}{4 \sqrt{\ln(2)}} \left(\frac{\theta_{Sa}^2}{\theta_{Sa}^2 + \theta_B^2} \right)^{1/2} \left(\frac{\theta_{Sb}^2}{\theta_{Sb}^2 + \theta_B^2} \right)^{1/2}} \\ &= \frac{1}{b \sqrt{\ln(2)}} \left(\frac{\theta_{Sa}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \left(\frac{\theta_{Sb}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2}. \end{aligned} \quad (16)$$

Inserting this value for $\frac{\Omega_s}{\Omega'_s}$ and using $\Omega_m = \frac{\pi b^2 \theta_B^2}{4}$ in Equation 10 yields

$$\begin{aligned} S_\nu(Jy) &= \frac{2k\nu^2}{c^2} \frac{\pi b^2 \theta_B^2}{4} \left(\frac{\theta_{Sa}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \left(\frac{\theta_{Sb}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \frac{T_A}{\eta_{mb}} \\ &= \frac{2k}{c^2} 10^{23} \left[\frac{(2\pi)(10^9)}{(360)(3600)} \right]^2 \frac{\pi b^2 \nu^2 (GHz) \theta_B^2 (arcsec)}{4} \\ &\quad \left(\frac{\theta_{Sa}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \left(\frac{\theta_{Sb}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \frac{T_A}{\eta_{mb}} (K) \\ &= 5.669 \times 10^{-7} b^2 \nu^2 (GHz) \theta_B^2 (arcsec) \\ &\quad \left(\frac{\theta_{Sa}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \left(\frac{\theta_{Sb}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \frac{T_A}{\eta_{mb}} (K) \\ S_\nu^{gauss}(Jy) &\simeq 0.750 \left(\frac{b\nu(GHz)}{115} \right)^2 \frac{\theta_B^2 (arcsec)}{100} \\ &\quad \left(\frac{\theta_{Sa}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \left(\frac{\theta_{Sb}^2 + \theta_B^2}{\theta_B^2} \right)^{1/2} \frac{T_A}{\eta_{mb}} (K). \end{aligned} \quad (17)$$

Note that for a point source $\theta_{Sa}, \theta_{Sb} \ll \theta_B$, and Equation 17 becomes:

$$\begin{aligned} S_\nu(Jy) &= \frac{2k\nu^2}{c^2} \frac{\pi b^2 \theta_B^2}{4} \frac{T_A}{\eta_{mb}} \\ &= \frac{2k}{c^2} 10^{23} \left[\frac{(2\pi)(10^9)}{(360)(3600)} \right]^2 \frac{\pi b^2 \nu^2 (GHz) \theta_B^2 (arcsec)}{4} \frac{T_A}{\eta_{mb}} (K) \\ &= 5.669 \times 10^{-7} b^2 \nu^2 (GHz) \theta_B^2 (arcsec) \frac{T_A}{\eta_{mb}} (K) \\ S_\nu^{point}(Jy) &\simeq 0.750 \left(\frac{b\nu(GHz)}{115} \right)^2 \frac{\theta_B^2 (arcsec)}{100} \frac{T_A}{\eta_{mb}} (K). \end{aligned} \quad (18)$$

Still assuming a point source, If you prefer to use the reflector diameter (D) rather than the beam size (θ_B), Equation 18 can be written as follows

$$\begin{aligned}
S_\nu^{point}(Jy) &= \frac{2k\nu^2}{c^2} \frac{\pi b^2}{4} \left(\frac{cb}{\nu D} \right)^2 \frac{T_A}{\eta_{mb}} \\
&= \frac{\pi k b^4}{2D^2} \frac{T_A}{\eta_{mb}} \\
&= \frac{\pi k b^4}{2D^2(m)} \frac{10^{23}}{(100)^2} \frac{T_A}{\eta_{mb}} (K) \\
&\simeq 2168.718 \frac{b^4 T_A}{D^2(m) \eta_{mb}} (K).
\end{aligned} \tag{19}$$

4 Uniform Disk Source

For a uniform disk source, such as a planet or asteroid, with equatorial and poloidal angular sizes θ_{eq} and θ_{pol} , and whose area, for a source radius R_s , is given by $\int_0^{R_s} 2\pi r dr$:

$$\begin{aligned}
\Omega'_S &= \int_0^{R_s} 2\pi r \left[\exp\left(-\frac{4r^2}{b^2\theta_B^2}\right) \right] dr \\
&= \frac{\pi b^2 \theta_B^2}{4} \left[1 - \exp\left(-\frac{\theta_{eq}\theta_{pol}}{b^2\theta_B^2}\right) \right],
\end{aligned} \tag{20}$$

where I have used the fact that

$$\frac{d}{dr} \left[\exp\left(-\frac{4r^2}{b^2\theta_B^2}\right) \right] = -\frac{8r}{b^2\theta_B^2} \left[\exp\left(-\frac{4r^2}{b^2\theta_B^2}\right) \right]. \tag{21}$$

Since the integral over the source is given by $\Omega_s = \int_0^{R_s} 2\pi r = \pi R_s^2$, we can write the source correction factor defining the coupling between a disk source and a Gaussian beam as follows

$$\begin{aligned}
\frac{\Omega_s}{\Omega'_s} &= \frac{\pi R_s^2}{\frac{\pi b^2 \theta_B^2}{4}} \left[1 - \exp\left(-\frac{\theta_{eq}\theta_{pol}}{b^2\theta_B^2}\right) \right]^{-1} \\
&= \frac{\theta_{eq}\theta_{pol}}{b^2\theta_B^2} \left[1 - \exp\left(-\frac{\theta_{eq}\theta_{pol}}{b^2\theta_B^2}\right) \right]^{-1}.
\end{aligned} \tag{22}$$

Inserting this value for $\frac{\Omega_s}{\Omega'_s}$ and using $\Omega_m = \frac{\pi b^2 \theta_B^2}{4}$ in Equation 10 yields

$$\begin{aligned}
S_\nu(Jy) &= \frac{2k\nu^2}{c^2} \frac{\pi b^2 \theta_B^2}{4} \frac{1}{b^2} \frac{\theta_{eq} \theta_{pol}}{\theta_B^2} \\
&= \left[1 - \exp\left(-\frac{\theta_{eq} \theta_{pol}}{b^2 \theta_B^2}\right) \right]^{-1} \frac{T_A}{\eta_{mb}} \\
&= \frac{2k}{c^2} 10^{23} \left[\frac{(2\pi)(10^9)}{(360)(3600)} \right]^2 \frac{\pi \nu^2 (GHz) \theta_{eq}(\text{arcsec}) \theta_{pol}(\text{arcsec})}{4} \\
&= \left[1 - \exp\left(-\frac{\theta_{eq} \theta_{pol}}{b^2 \theta_B^2}\right) \right]^{-1} \frac{T_A}{\eta_{mb}} (K) \\
&= 5.669 \times 10^{-7} \nu^2 (GHz) \theta_{eq}(\text{arcsec}) \theta_{pol}(\text{arcsec}) \\
&= \left[1 - \exp\left(-\frac{\theta_{eq} \theta_{pol}}{b^2 \theta_B^2}\right) \right]^{-1} \frac{T_A}{\eta_{mb}} (K) \\
S_\nu^{disk}(Jy) &\simeq 0.749 \left(\frac{\nu(GHz)}{115} \right)^2 \frac{\theta_{eq}(\text{arcsec}) \theta_{pol}(\text{arcsec})}{100} \\
&= \left[1 - \exp\left(-\frac{\theta_{eq} \theta_{pol}}{b^2 \theta_B^2}\right) \right]^{-1} \frac{T_A}{\eta_{mb}} (K). \tag{23}
\end{aligned}$$

Note that for a point source $\theta_{eq}, \theta_{pol} \ll \theta_B$, and Equation 23 becomes:

$$\begin{aligned}
S_\nu(Jy) &= \frac{2k\nu^2}{c^2} \frac{\pi b^2 \theta_B^2}{4} \frac{T_A}{\eta_{mb}} \\
&= \frac{2k}{c^2} 10^{23} \left[\frac{(2\pi)(10^9)}{(360)(3600)} \right]^2 \frac{\pi b^2 \nu^2 (GHz) \theta_B^2(\text{arcsec})}{4} \frac{T_A}{\eta_{mb}} (K) \\
&= 5.669 \times 10^{-7} b^2 \nu^2 (GHz) \theta_B^2(\text{arcsec}) \frac{T_A}{\eta_{mb}} (K) \\
S_\nu^{point}(Jy) &\simeq 0.750 \left(\frac{b\nu(GHz)}{115} \right)^2 \frac{\theta_B^2(\text{arcsec})}{100} \frac{T_A}{\eta_{mb}} (K), \tag{24}
\end{aligned}$$

which is exactly the same relation derived from the assumption of an elliptical gaussian source and beam in Equation 18 (as it should be).