### **Supplementary Information for**

# **Strong and tunable spin-orbit interaction in a single crystalline InSb nanosheet**

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### **Supplementary Note I. Determination of the carrier density and carrier mobility in the InSb nanosheet**



Figure 1. (a) Conductance *G* (black line) measured for the InSb nanosheet in the dualgate device as a function of back gate voltage  $V_{BG}$  at top gate voltage  $V_{TG} = 0$  V and at  $T = 1.9$  K. The red line is a linear fit to the measurements. This fitting line is extended to intersect the back-gate axis at which the back gate threshold voltage  $V_{BG}^{th} = -15$  V (as marked by a yellow arrow) is extracted. (b) Conductance *G* (black line) measured for the InSb nanosheet in the dual-gate device as a function of top gate voltage  $V_{TG}$  at bottom gate voltage  $V_{BG} = 0$  V and at  $T = 1.9$  K. The red line is a linear fit to the measurements. This fitting line is extended to intersect the top-gate axis at which the top gate threshold voltage  $V_{TG}^{th} = -0.75$  V (as marked by a yellow arrow) is extracted.

In this Supplementary Note, we describe how the carrier density and carrier mobility in the InSb nanosheet of the dual-gate device are extracted from the measured gate transfer characteristics of the device. Figure 1(a) shows the conductance of the nanosheet measured at  $T = 1.9$  K as a function of the back gate voltage  $V_{BG}$  (back gate transfer characteristics) with top gate voltage fixed at  $V_{TG} = 0$  V. The measurements are carried out in a four-probe configuration (see Fig. 1a in the main article for the measurements circuit setup), in which a 17 Hz AC excitation current *I* of 100 nA is applied through the two outer electrodes and the voltage drop *V* between the two inner electrodesis reordered. Because the effect of the contact resistances has been eliminated in such four-probe measurements, the conductance of the nanosheet can be obtained directly from  $G = I/V$ . The carrier density in the nanosheet can be estimated from  $n = C_{gs} \times \frac{V_{BG} - V_{BG}^{th}}{e}$ , where  $C_{gs}$  is the unit area capacitance of the back gate and  $V_{BG}^{th}$ is the back gate threshold voltage at which the conductance of the nanosheet goes to zero. Here we estimate  $C_{gs}$  using a parallel capacitor model  $C_{gs} = \frac{\varepsilon \varepsilon_0}{d}$ , where  $\varepsilon_0$  is the vacuum permittivity,  $\varepsilon$  and  $d$  are the dielectric constant and thickness of the gate dielectric. Using  $\varepsilon = 3.9$  and  $d = 300$  nm for the dielectric layer of SiO<sub>2</sub> in this work, we obtain  $C_{gs} = 1.15 \times 10^{-8}$  F · cm<sup>-2</sup>. The back gate threshold voltage  $V_{BG}^{th}$  is extracted from the measured back-gate transfer characteristics shown in Fig. 1(a) as follows. First, we fit the measured back-gate transfer characteristics by a line (red line in the figure). Then, we extend the fitting line to intersect the back-gate voltage axis and the back gate value at the intercept is the extracted value for  $V_{BG}^{th}$ . From the measurements shown in Fig. 1(a),  $V_{BG}^{th} \sim -15$  V is obtained. The carrier density in the nanosheet can now be evaluated at a given value of  $V_{BG}$ . For example, at  $V_{BG} = -5$  V (corresponding to a case with the nanosheet channel conductance of  $G \sim 9e^2/h$ ), a carrier density of  $n = 7.2 \times 10^{11}$  cm<sup>-2</sup> in the nanosheet is obtained, while at  $V_{BG}$  = −9 V (corresponding to a case with the nanosheet channel conductance of  $G \sim 5 e^2/h$ , a carrier density of  $n = 4.3 \times 10^{11}$  cm<sup>-2</sup> in the nanosheet is obtained.

The carrier mobility in the InSb nanosheet can be obtained from  $\mu = \sigma/ne$ , where  $\sigma = \frac{GL}{W}$  is the nanosheet channel conductivity with L being the channel length (i.e., the distance between the two inner contacts, about 1.1  $\mu$ m in this device) and W the channel width (i.e., the width of the nanosheet, about 550 nm in this device). Here, we note that since both the nanosheet conductance and the carrier density in the

nanosheet depend linearly on  $V_{BG}$ , the extracted carrier mobility from the transfer characteristic measurements (which is often called the field effect mobility) will be independent of  $V_{BG}$ . Thus, we can evaluate the carrier mobility  $\mu$  by setting the back gate voltage value at, e.g.,  $V_{BG} = -5$  V, at which the carrier density is  $n = 7.2 \times$  $10^{11}$  cm<sup>-2</sup> and the nanosheet conductance is  $G \sim 9e^2/h$ . The obtained carrier mobility is then  $\mu \sim 6000 \text{ cm}^2 \cdot V^{-1} \cdot s^{-1}$ . The carrier mean free path in the nanosheet is given by  $L_e = \frac{\hbar \mu}{e} \sqrt{2\pi n}$ , where  $\hbar = \frac{\hbar}{2\pi}$  with  $h$  being the Planck constant. From the measured back gate transfer characteristics shown in Fig. 1(a), we obtain  $L_e \sim 84$  nm at  $n = 7.2 \times 10^{11}$  cm<sup>-2</sup> (and  $G \sim 9e^2/h$ ) and  $L_e \sim 65$  nm at  $n =$  $4.3 \times 10^{11}$  cm<sup>-2</sup> (and  $G \sim 5e^2/h$ ).

Using the top-gate transfer characteristics shown in Fig. 1(e), similar results for the carrier density and electron mobility in the InSb nanosheet should be extracted. However, since the dielectric constant of  $HfO<sub>2</sub>$  in our device is an unknown parameter, which has been given to over a wide range of values in the literature, a direct estimation of the carrier density and electron mobility in the InSb nanosheet from the top-gate transfer characteristics is not possible. Nevertheless, using the results obtained above, we can determine the dielectric constant of  $HfO<sub>2</sub>$  employed in our device. The equation to be used for extraction of the carrier density based on the top-gate transfer characteristics becomes  $n = C_{gs} \times \frac{V_{TG} - V_{TG}^{th}}{e}$ , where  $C_{gs} = \frac{\varepsilon \varepsilon_0}{d}$  with  $\varepsilon$  being the unknown dielectric constant of HfO<sub>2</sub>,  $d = 20$  nm the thickness of HfO<sub>2</sub>, and  $V_{TG}^{th}$  the top-gate threshold voltage at which the conductance *G* goes to zero. Similarly as in Fig. 1(a),  $V_{\text{TG}}^{\text{th}}$  can be extracted from the measurements shown in Fig. 1(b) by a linear fit to the low top-gate voltage data and by extending the fitting line to intersect the top-gate voltage axis. As seen in Fig. 1(b), a result of  $V_{\text{TG}}^{\text{th}} \sim -0.75$  V is obtained. To determine the dielectric constant  $\varepsilon$  of HfO<sub>2</sub> in our device, we consider the case of the conductance  $G \sim 9e^2/h$  at  $V_{TG} = -0.35$  V, corresponding to the case of carrier density *n* =  $7.2 \times 10^{11}$  cm<sup>-2</sup> as estimated through the bottom-gate transfer characteristics. By taking this value into the above equation, a value of  $\varepsilon \sim 6.5$  can be obtained for the dielectric constant of  $HfO<sub>2</sub>$  in our device. This value indicates that the HfO2 layer in our device is in good amorphous phase, consistent with the fact that it was grown at a low temperature by atomic layer deposition.

## **Supplementary Note II. Comparison between the results obtained by analyses of magnetotransport measurements of the InSb nanosheet using the HLN and ILP theories**

In the main article, the HLN model is utilized in the analysis of our magnetotransport data. This is suitable for a weak disordered system such as InSb nanosheet and other emerging 2D materials, where the electron elastic scattering length, or the mean free path,  $L_e$  is shorter than all other characteristic transport length scales, such as phase coherence length  $L_{\varphi}$  and spin-orbit length  $L_{\text{SO}}$ . However, in a clean 2D electron system with a ultrahigh mobility made from a semiconductor heterostructured quantum well,  $L_e$  can be exceedingly longer than  $L_{SO}$ . In this case, the HLN model may no longer be applicable and one might need to invoke the so-called ILP model, developed by Iordanskii, Lyanda-Geller and Pikus<sup>1</sup>, in analyses of the magnetotransport measurement data. Here, it is worthwhile to check whether the ILP model can be applied to the magnetotransport data obtained in our device. In the ILP model, the quantum conductance correction to the low-field magnetoconductance is given by

$$
\Delta \sigma_{\rm ILP} = -\frac{e^2}{4\pi^2 \hbar} \left\{ \frac{1}{a_0} + \frac{2a_0 + 1 + \frac{H_{\rm SO}}{B}}{a_1 \left[ a_0 + \frac{H_{\rm SO}}{B} \right] - 2\frac{H_{\rm SO}}{B}} + 2\ln\frac{H_{\rm tr}}{B} + \psi \left( \frac{1}{2} + \frac{H_{\phi}}{B} \right) + 3C - \sum_{n=1}^{\infty} \left[ \frac{3}{n} - \frac{3a_n^2 + 2a_n \frac{H_{\rm SO}}{B} - 1 - 2(2n+1)\frac{H_{\rm SO}}{B}}{[a_n + \frac{H_{\rm SO}}{B}]a_{n-1}a_{n+1} - 2\frac{H_{\rm SO}}{B}[(2n+1)a_n - 1]} \right] \right\}.
$$
\n(1)

Here,  $H_{\text{SO}} = H'_{\text{SO}} + H_{\text{SO}}^3$  with  $H'_{\text{SO}}$  being Rashba term and  $H_{\text{SO}}^3 > 0$  the cubic Dresselhaus term,  $a_n = n + \frac{1}{2} + \frac{H_{\varphi}}{B}$  $\frac{H_{\phi}}{B} + \frac{H_{\text{SO}}}{B}$ , *C* is Euler's constant,  $H_{\text{tr}} = \frac{\hbar}{4eL_{\text{e}}^2}$ , and  $H_{\varphi} = \frac{\hbar}{4eL_{\varphi}^2}$ . In the calculations using the above equation, the summation of first 10000 terms in the series has been performed and it is checked that a desired convergence has been achieved.

Figure 2 given below shows a comparison between the results of analyses using the HLN and ILP theories, where the magnetoconductance data (dots) measured for our InSb nanosheet device at  $T = 1.9$  K,  $V_{TG} = -0.46$  V and  $V_{BG} = 1.54$  V, and the best fits to the data by both the HLN (light blue line) and ILP (light green line) models are presented. Clearly, the HLN model yields a satisfactory fit to the measurement data, giving the extracted length scales of  $L_{\varphi}$  = 472 nm,  $L_{\text{SO}}$ =137 nm and  $L_{\varphi}$  = 88 nm. However, large deviations from the measurements are found in the best fit to the ILP model. In addition, the best fit by the ILP model gives  $H'_{so} = 4.35 \times 10^{-2}$  T,  $H_{\varphi} =$ 

 $4.04 \times 10^{-4}$  T,  $H_{tr} = 2.98 \times 10^{-2}$  T, and  $H_{so}^3 = 6.2 \times 10^{-21}$  T ( $\approx 0$  T), and thus the length scales of  $L_{\varphi}$  = 638 nm,  $L_{\text{SO}}$  = 61 nm and  $L_{\text{e}}$  = 74 nm. Here, both the values of  $L_{\varphi}$  and  $L_{\varphi}$  may look reasonable, but the value of  $L_{\text{so}} \left( \langle L_{\varphi} \rangle \right)$  looks unphysical for the InSb nanosheet in our device. Thus, for our InSb nanosheet device, it is more appropriate to use the HLN model, instead of the ILP model, for analyses of our measured magnetoconductance data.



Figure 2. Comparison between the best fits of the magnetoconductance data measured for our InSb nanosheet device at  $T = 1.9$  K,  $V_{TG} = -0.46$  V and  $V_{BG} = 1.54$  V to the HLN and ILP models.

### **Supplementary Note III. Materials parameters and simulations for energy band diagrams**

To simulate the energy band diagrams of the  $HfO_2$ -InSb-Si $O_2$  structure in the dualgate InSb nanosheet device, Poisson's equation is solved using commercially available program COMSOL in compliance with the boundary conditions set in our experiment. Here we assume that each material layer is an infinite two-dimensional structure and we thus need to solve effectively only a one-dimensional Poisson's equation. Material parameters of InSb,  $SiO<sub>2</sub>$  and  $HfO<sub>2</sub>$  utilized in the simulations for the energy band diagrams are given in Table I.

Poisson's equation used here to describe the electrostatics of the  $HfO_2$ -InSb-SiO<sub>2</sub> heterostructure has a form of

$$
\nabla \cdot (-\epsilon_{\mathbf{r}} \nabla V) = q(p - n + N_{\mathbf{d}}^+ - N_{\mathbf{a}}^-),\tag{2}
$$

where V is the electric potential,  $\epsilon_r$  is the dieletric constant of the material, q is the

elementary charge, p and n are the hole and electron densities,  $N_d^+$  and  $N_a^-$  are the ionized donor and acceptor concentrations, respectively. The energies of the conduction and valence band edges can be calculated as,  $E_c = -(eV + \chi_0)$  and  $E_V = -(eV + \chi_0)$  $\chi_0 + E_{g,0}$ , where  $\chi_0$  and  $E_{g,0}$  are the electron affinity and bandgap of a material. The continuity conditions at the interface of two different materials are,  $\mathbf{n} \cdot (D_1 - D_2) = 0$ and  $E_{F,1} = E_{F,2}$ , where **n** denotes the normal vector of the interface,  $D_{1,2}$  and  $E_{F,1,2}$  are the electric displacements and electron Fermi levels in the two materials.

Table I. Material parameters of InSb,  $SiO<sub>2</sub>$  and HfO<sub>2</sub> utilized in the simulations for the HfO<sub>2</sub>-InSb-SiO<sub>2</sub> heterostructure in the dual-gate InSb nanosheet device at  $T = 2$  K.

Material	Bandgap (eV)	Dielectric Constant	<b>Electron Effective</b> Mass $(m_0)$	Electron Affinity (eV)
<b>InSb</b>	$0.23^{(1)}$	16.8	0.014	$4\,77^{(2)}$
HfO <sub>2</sub>	5.8	$6.5^{(3)}$	0.11	$2.8^{(4)}$
SiO <sub>2</sub>	8.95	3.9	0.3	$0.75^{(5)}$

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Figure 3(a) is the zoom-in figure of Fig. 4b in the main article which displays the calculated profiles of the conduction band edge inside the InSb layer at the experimental condition of the carrier density  $n = 7.2 \times 10^{11}$  cm<sup>-2</sup> and the conductance  $G \sim 9e^2/h$ with three different values of voltage  $V_D$  applied over the dual gate. Clearly the conduction band edge inside the InSb layer is bended, leading to the presence of an electric field in the layer. Figure 3(b) shows the calculated electric field distribution inside the InSb layer at the three considered values of  $V<sub>D</sub>$ . Clearly, among the three cases, an overall strongest field strength is found inside the InSb layer for  $V_D = -2$  V, which should give a strong SOI. On the contrast, at  $V_D = 11$  V, the field strength inside the InSb layer is overall small, which should produce a relatively weak SOI. Figure 3(c) displays the calculated carrier density distributions in the InSb layer at the three values of  $V<sub>D</sub>$ . As shown in Fig. 3(c), at  $V<sub>D</sub> = -2$  and 0 V, carriers are non-uniformly distributed and they mainly concentrated to the bottom part of the InSb layer, close to the  $SiO<sub>2</sub>$ dielectric, where strong electric fields are presented and carriers will experience a

strong Rashba SOI when they move along the layer. At  $V_{\text{D}} = 11$  V, the carrier distribution becomes less non-uniform in the InSb layer with a significant amount appearing in the middle region of the layer, where the electric field is comparably weak and the carriers would experience a weak SOI in the InSb nanosheet. All these simulated results are in a good agreement with our experimentally measured results for the SOI in the InSb nanosheet.



Figure 3. (a) Calculated conduction band edges (red, blue and green solid lines) and Fermi levels (red, blue and green dashed lines) inside the InSb layer with a sheet carrier density  $n = 7.2 \times 10^{11}$  cm<sup>-2</sup> at three different values of dual-gate voltage  $V_D$ . (b) Calculated corresponding effective electric field strengths inside the InSb layer at the same three values of  $V_D$  as in (a). (c) Calculated corresponding carrier density distributions inside the InSb layer at the same three values of  $V_D$  as in (a). The inset shows the results of the calculations at  $V_{\text{D}} = -2 \text{ V}$  with the carrier density plotted in logarithmic scale.

#### **Supplementary Note IV. Analysis of the Rashba SOI in the InSb nanosheeet**

In a semiconductor quantum structure, two predominant mechanisms that give rise to spin-orbit coupling and thus lift the spin degeneracy even in the absence of a magnetic field are the Dresselhaus and Rashba<sup>2</sup> SOI. The first one arises from an intrinsic bulk inversion asymmetry (BIA) of the underlying crystal structure, as described by Dresselhaus<sup>3</sup>, while the second one arises from a structural inversion asymmetry (SIA) induced by an electrical field  $\mathbf{E} = -\nabla V(\mathbf{r})$  in the crystal, where  $V(\mathbf{r})$  is the electric potential, as described by Bychkov and Rashba<sup>4</sup>. The electric field could include both a built-in part in the structure and a tunable part created by, e.g., applying a gate voltage.

In the lowest-order approximation, the Rashba SOI Hamiltonian can be written as<sup>5</sup>

$$
H_{\rm R} = r_{\rm R} \sigma \cdot \mathbf{k} \times \mathbf{E},\tag{3}
$$

where  $r_R$  is a material-specific, Fermi level dependent prefactor<sup>6,7</sup> and **k** is the wave vector. To estimate the effect of the Rashba SOI in our dual-gate device structure, we assume that all conduction carriers experience a same electric field in the InSb nanolayer. We approximate this electric field by the mean electric field  $\mathbf{E} = (0, 0, E)$ , with  $E$  being the perpendicular component of the electrical field obtained by averaging through the InSb nanolayer along the perpendicular direction. The wave vector only has in-plane components and can be written as  $\mathbf{k} = (k_x, k_y, 0)$ . Rashba Hamiltonian then becomes  $H_R = \alpha_R (k_y \sigma_x - k_x \sigma_y)$ , where  $\alpha_R = r_R E$  is known as the Rashba SOI strength parameter. Moreover, the spin-orbit precession length is given by  $L_{\rm R} = \frac{\hbar^2}{m^* \alpha_{\rm R}}$  with  $m^*$  being the elecron effective mass. Therefore  $\frac{1}{L_{\rm R}}$  is in proportion to  $\alpha_R$  and thus in proportion to the mean electric field  $E$ .

The extracted spin-orbit relaxation length  $L_{SO}$  from the measured magnetoconductance data in our experiments comprise contributions from comprehensive spin relaxation processes induced by all kinds of SOIs (i.e., Rashba SOI, Dresselhaus SOI and other high-order kinds). It is naturally a hypothesis that the Rashba SOI induced spin precession process is the major cause for the WAL characteristics observed in our magnetoconductance measurements. Thus, the spin procession induced relaxation time  $\tau_{SO}$  caused by all SOIs could be written as  $\frac{1}{\tau_{SO}} = \frac{1}{\tau_R} + \cdots$ , where  $\tau_R$ is the spin relaxation time cause by the Rashba SOI. As a consequence, we have  $\frac{1}{L_{\text{SO}}^2}$  =  $\frac{1}{R^2} + \cdots$  and thus expect to see that  $\frac{1}{L_{\text{SO}}^2} = \left(\frac{m^* r_R}{\hbar^2}\right)$  $\int_{0}^{2} E^{2} + C_{0}$  in the InSb nanolayer in our device, where  $C_0$  is a constant by assuming that the Elliot-Yafet term, Dresselhaus SOI term and all other high-order terms are electric field independent. To see this, we plot in Fig. 4, the extracted  $\frac{1}{L_{\text{SO}}^2}$  as a function of  $E^2$  at carrier densities of  $n = 7.2 \times$  $10^{11}$  cm<sup>-2</sup> and  $4.3 \times 10^{11}$  cm<sup>-2</sup>. The red and blue dashed lines in the figure show the linear fits to the data at the two different carrier densities, namely different Fermi levels. As shown in Fig. 4, at both carrier densities,  $\frac{1}{L_{SO}^2}$  displays a good linear dependence on  $E^2$ . The slopes  $\kappa = \left(\frac{m^* r_R}{\hbar^2}\right)$ of the fitting lines are 0.613 V<sup>-2</sup> at  $n = 7.2 \times 10^{11}$ 

gate voltage tunable WAL characteristics. The intercepts of the two fitting lines and the vertical axis are nearly the same and are very close to a value of  $8 \mu m^2$ , which gives  $C_0 = 8 \mu m^{-2}$  and represents all other field-independent contributions including the one from the Dresselhaus SOI. The Fermi level dependent prefactors  $r_R$  can be obtained from the extracted slopes *κ* using the relation  $r_R = \frac{\hbar^2}{m^*} \cdot \sqrt{\kappa}$ . The results are  $r_R = 4.26$  $e \cdot \text{nm}^2$  at  $n = 7.2 \times 10^{11}$  cm<sup>-2</sup> and  $r_R = 3.48 e \cdot \text{nm}^2$  at  $n = 4.3 \times 10^{11}$  cm<sup>-2</sup>.



Figure 4. Extracted  $\frac{1}{l_{\text{SO}}^2}$  versus calculated mean field strength  $E^2$  in the InSb nanolayer of the dual-gate device. Red and blue dots are the data points obtained at sheet carrier densities  $n = 7.2 \times 10^{11}$  cm<sup>-2</sup> and  $4.3 \times 10^{11}$  cm<sup>-2</sup>, respectively. Red and blue dashed lines are the linear fits to the data. The two fitting lines intersect the vertical axis at nearly the same value of  $C_0 \sim 8 \mu m^2$  as marked by a yellow arrow.

## **Supplementary Note V. Dual-gate voltage dependent measurements of the magnetoconductance along the constant conductance contour lines of ~1.1 and**   $-2.6 e^2/h$

The dual-gate voltage dependent measurements of the magnetoconductance and the characteristics transport lengths have also been performed for the InSb nanosheet at two lower carrier densities, i.e., along the constant conductance contour lines of ~1.1 and  $\sim$ 2.6  $e^2/h$ . Figure 5 summarizes the measurements, where Figs. 5(a) and 5(b) show the results along the constant conductance contour line of  $\sim$ 2.6  $e^2/h$ , while Figs.  $5(c)$  and  $5(d)$  show the results along the constant conductance contour line of  $\sim 1.1$  $e^2/h$ . It is seen that similar dual-gate voltage dependences of the transport lengths  $L_{\omega}$ ,  $L_{\rm so}$ , and  $L_{\rm e}$  as observed in Fig. 3 of the main article are obtained. In particular, the spin-orbit length  $L_{\text{SO}}$  is seen to be efficiently controlled via the dual-gate voltage  $V_{\text{D}}$ at both constant conductance values. These data, together with those shown in Fig. 3 of the main article, demonstrate that the SOI in the InSb nanosheet in a dual-gate structure can be efficiently tuned by a voltage applied to the dual gate at largely different but fixed carrier densities of the nanosheet.



Figure 5. (a) Low-field magnetoconductance  $\Delta G$  measured for the device at a constant conductance value of  $\sim 2.6 \, e^2/h$  and temperature  $T = 1.9$  K at various values of the voltage  $V_D = V_{TG} - V_{BG}$  applied over the dual gate. The bottom trace shows the magnetoconductance data measured at  $V_D = -11.3$  V and all other measured magnetoconductance traces are successively vertically offset for clarity. The black solid lines are the theoretical fits of the experimental data to the HLN equation [Eq. (1) in the main article]. (b) Phase coherence length  $L_{\varphi}$ , spin-orbit length  $L_{\text{SO}}$ , and mean free path  $L_e$  extracted from the fits in (a) as a function of  $V_D$ . (c) The same as (a) but measured for the device at a constant conductance value of ~1.1  $e^2/h$ . Here, the bottom trace shows the magnetoconductance data measured at  $V_D = -4.9$  V and all other measured magnetoconductance traces are again successively vertically offset for clarity. (d) Phase coherence length  $L_{\varphi}$ , spin-orbit length  $L_{\text{SO}}$ , and mean free path  $L_{\text{e}}$ extracted from the fits in (c) as a function of  $V<sub>D</sub>$ .

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