Cooperative Communication Technologies for LTE-Advanced

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Abstract—The LTE-Advanced (LTE-A) system is currently under development to allow for significantly higher spectral efficiency and data throughput than LTE systems. In a wireless system based on orthogonal frequency division multiplexing (OFDM) with frequency reuse factor one such as LTE, the achievable cell spectral efficiency is often limited by the inter-cell interference or coverage shortage of base stations. Hence in LTE-A, coordinated multi-point (CoMP) transmission/reception (a.k.a. multi-cell MIMO or base station cooperation) and relaying technologies are being introduced to clear these major performance hurdles. In this paper, overall picture of cooperative communication technologies being discussed in LTE-A systems including CoMP and relaying is presented, together with considerations on system design.

Index Terms—relays, cooperative systems

I. INTRODUCTION

Next generation wireless communication systems, named IMT-Advanced (IMT-A) systems, target to achieve another major advance from the current 3G system, in terms of achieving 1Gbps for downlink (DL) and 500 Mbps for uplink (UL) throughput. For this purpose, 3GPP society is currently developing LTE-Advanced (LTE-A) standard [1] as an evolution of the already frozen LTE standard [2].

From information theory we know that the ultimate performance measure of a communication system is the spectral efficiency [3]. Furthermore, the spectral efficiency of a communication link is determined by signal-to-noise-plusinterference ratio (SINR) at a receiver. To be specific, the SINR at a receiver can be written as

$$SINR = \frac{P}{I+N}$$

where P is power seen at the receiver of a signal transmitted by a transmitter, I is interference power from other interfering sources and N is variance of an additive white Gaussian noise signal.

In most cases, a low SINR is caused by either of two scenarios: *noise-limited* scenario and *interference-limited* scenario [4].

In the noise-limited scenario, we have

$$I << P$$
 and $P \sim N$.

This implies that interference is negligible compared to noise variance and a low SINR is mainly caused by the fact that the received signal strength of the targeted signal, P, is low. A natural solution to boost the SINR is to increase P utilizing a relaying technology.

Two classes of relaying schemes considered in LTE-A are studied in this paper: type I and type II. In the type I relaying scheme, also known as self-backhauling, a relay is seen as a serving base station at noise-limited mobile terminals, and it can forward Internet protocol (IP) packets to these mobile terminals and to a base station. In the type II relaying scheme, also called as transparent relaying for HARQ, a relay is deployed to help hybrid ARQ retransmissions, which may also improve cell-edge terminals' throughput.

On the other hand, in the *interference-limited* scenario, we have,

N << I and $I \sim P$.

In this case, noise power is negligible compared to interference power and a low SINR is mainly due to the fact that interference power I is large. Then, a natural solution to improve SINR is to apply cooperative communication techniques to turn interference to useful signals.

In OFDM based systems, interference is mainly originated from the other base stations' (or cells') transmitted signals for serving their own mobile terminals (called inter-cell interference). Therefore, base station cooperation ([5], for example), also known as coordinated multi-point (CoMP) transmission or multi-cell MIMO, can be one of the most important enabling technologies beneficial in scenarios with high inter-cell interference in the next generation wireless communication systems. Two major categories of CoMP schemes discussed in LTE-A are considered in this paper: joint processing and coordinated beamforming/scheduling. In joint processing, multiple base stations jointly transmit signals to a mobile terminal, whereby signals from other cells turn from intercell interference into useful signals. In coordinated beamforming/scheduling, base stations either jointly choose precoding matrices, not only matching to their own serving terminals' channels but also less interfering to terminals scheduled in adjacent cells.

II. RELAYING TECHNOLOGIES

Relaying has been extensively studied in information theory, communication theory and signal processing societies. Mainly two categories of relaying have been considered. In one category, individual nodes in a wireless network having their own data packets to send, cooperate each other to increase the network throughput by relaying signals or packets of other nodes (e.g., [6], [7], [8]). In another category, in a wireless

network, altruistic relay nodes that do not have their own packets to send are deployed to help other nodes' transmissions (e.g., [9], [10]), where inter-user cooperation is not allowed. In commercial cellular wireless applications, the main focus of relaying discussions is on the second category, partly because of the concerns on spending each mobile terminal's resources for purposes other than its own benefit.

In the context of cellular networks with spectral reuse factor 1, introduction of relays result in increased interference to mobile terminals, when a relay and a base station are not allowed to cooperate. This is because the same time-frequency resource could be used by a relay and a base station serving their respective mobile terminals. One obvious solution to solve this problem is to allow cooperative transmissions between a base station and a relay or among relays. However, this in practice involves huge control signaling overhead that reduces useful wireless bandwidth for data transmission: the overhead could comprise signaling supports for synchronized scheduling and sharing channel state information for link adaptation (i.e., control of modulation size and coding rate adapting to channel state), and so on. Hence in LTE-A, the scope of relaying is mainly confined in coverage-limited case, to help mobile terminals located a place where a base station's power is not sufficiently received for successful demodulations.

For coverage extension, LTE-A considers mainly three relaying technologies, repeater relaying (also known as amplifyand-forward), self backhauling and transparent relaying for HARQ retransmissions. Since repeater or amplify-and-forward has been throughly studied in the literature (for example, [9], [10]), in this section, we focus on the two remaining relaying schemes, i.e., self backhauling and transparent relaying for HARQ retransmissions.

In the sequel, we refer to a link between a base station and a relay node as a backhaul link, a link between a relay node and a mobile terminal as an access link, and a link between a base station and a mobile terminal as a macro-access link.

A. Self-Backhauling

Relays in self-backhauling relaying (or type-I relays) are decode-and-forward relays, which receive and forward IP packets in the network layer, where physical layer cooperation among relays and other nodes are not allowed. In this case, a relay has its own scheduler and HARQ control, and mobile terminals may perceive a relay as another base station. On one hand, this type-I relaying design greatly reduces burden of system design at least at the mobile terminal side, since mobile terminal does not need to distinguish a relay and a base station. Hence, it is possible to deploy relays in an existing wireless network without changing existing mobile terminals' behavior. On the other hand, as both the access link and the macro-access link can be activated in the same time-frequency resources without coordination, a relay's signal may interfere with a base station's signal received at a mobile terminal. In this case, similar considerations as needed for cell planning should be taken for deploying this type of relays for managing interference between relays and base stations. If no interference coordination is to be implemented between a relay and a base station, the deployment of a type-I relay should be restricted in coverage-limited scenarios, where placement of a relay does not incur large interference to mobile terminals served by a base station, as well as the placement of the relay helps reception of IP packets at mobile terminals to which signals from the base station is weak.

1) Performance Analysis: To see the performance of the type-I relaying in a closer look, we consider a four-node network composed of a base station, two mobile terminals and one type-I relay placed on a line. We assume that the base station is placed at position 0, the two mobile terminals at d_1 and d_2 , and the relay at d_3 , where $d_1 < d_2 < d_3$. We further assume that a channel gain between two nodes is determined by pathloss only, and the pathloss exponent is γ . In addition, the base station and the relay transmits signals with power P_1 and P_2 . Without the relay, the SINRs at the two mobile terminals are, respectively,

$$\operatorname{SINR}_1 = \frac{d_1^{-\gamma} P_1}{N} \quad \text{and} \quad \operatorname{SINR}_2 = \frac{d_2^{-\gamma} P_1}{N}. \tag{1}$$

With the introduction of the relay, we suppose that the base station transmits signals to mobile terminal 1 and the relay serves mobile terminal 2 in the same frequency band. In type-I relaying, the relay and the base station do not cooperate in the physical layer, and hence the relay's signal and the base station's signal interfere with each other at each mobile station. Then, the SINRs with the relay are,

$$\operatorname{SINR}_{1}^{\prime} = \frac{d_{1}^{-\gamma} P_{1}}{N + (d_{3} - d_{1})^{-\gamma} P_{2}} \text{ and } \operatorname{SINR}_{2}^{\prime} = \frac{(d_{2} - d_{3})^{-\gamma} P_{2}}{N + d_{2}^{-\gamma} P_{1}}$$

The network throughput values η_1 and η_2 in terms of bits/sec with and without the relay are, respectively,

$$\eta_1 = \log_2(1 + \text{SINR}_1) + \log_2(1 + \text{SINR}_2)$$
(3)

$$\eta_2 = \alpha (\log_2(1 + \text{SINR}_1') + \log_2(1 + \text{SINR}_2')), \quad (4)$$

where $\alpha \in (0, 1)$ is a fraction of time available for transmissions to mobile terminals. If $d_3 >> d_1$ and $d_2 - d_3 << d_2$, we have SINR'₁ \simeq SINR₁ and SINR'₂ > SINR₂, hence deploying a type-I relay could potentially give throughput gain in the network thanks to the extended coverage to mobile terminal 2. However, this is not always the case. Suppose that $d_2 = 3d_1$, $d_3 = 2d_1$, $P_2 = 0.1P_1$ and $\gamma = 3$. Then, we have SINR'₁ < SINR₁ and SINR'₂ \simeq SINR₂ and hence deploying a type-I relay decreases the network throughput owing to an increased interference to mobile terminal 1.

2) Design Challenges: Even though the motivation of adopting type-I relaying is mainly for facilitating a simple design, there are still a lot of issues to be resolved for seamless operation of LTE-A cellular networks and for ensuring backward compatibility to LTE users. Most of the challenges of designing type-I relaying in LTE-A comes from relay's being half-duplex owing to self interference, implying that an access link and a backhaul link associated with a relay cannot be activated simultaneously.

As a type-I relay is known as a base station at some mobile terminals, the type-I relay has to transmit first a few symbols in every subframe for control signaling and pilot signaling via an access link, to ensure backward compatibility to LTE. When control regions of a base station and a relay overlap in subframes, the relay is not able to receive control signaling from the base station due to being half-duplex. Hence, a new control signaling channels and methods backhaul link should be defined for the type-I relaying in backhaul links. On the other hand, in a subframe where a relay receives data from a base station in a backhaul link, the relay has to switch from transmitting mode for an access link for control signaling for its mobile terminals, to receiving mode for a backhaul link. This mode switching usually requires a time delay, this latency should be considered at a base station for designing subframes intended for transmitting data to relay in a backhaul link.

B. Transparent Relaying for HARQ retransmissions

As another attempt to extend coverage of LTE-A systems while minimizing impacts on the system design and overhead, a transparent relaying scheme for HARQ retransmissions (also known as type-II relaying) has been introduced. In type-II relaying, relays are activated only for HARQ retransmissions for mobile terminals. For example, a relay listens to downlink transmissions from a base station and uplink ACK/NACK messages from mobile terminals. When the relay decodes a NACK message from a mobile terminal which has decoded the associated packet to the NACK message, it can jointly transmit retransmission signals intended for the mobile terminal together with the base station.

1) Performance Analysis: For analysis of the type-II relaying, we consider a three-node network composed of a base station, a type-II relay and a mobile terminal. We assume that the first round error probability of a packet transmitted by the base station at the mobile terminal is p_1 , and the second round probability at the mobile terminal when the relay does not help is p_2 . Furthermore, we assume that the second round error probability at the mobile terminal when the relay helps is p'_2 . The throughput values η and η' in terms of packets per time slot, with and without the type-II relay can be calculated as,

$$\eta = (1 - p_1) + p_1(1 - p_2)/2 = 1 - 0.5p_1 - 0.5p_1p_2, \quad (5)$$

$$\eta' = (1 - p_1) + p_1(1 - p'_2)/2 = 1 - 0.5p_1 - 0.5p_1p'_2.$$
 (6)

Hence, deploying a type-II relay will have throughput gain if $p'_2 < p_2$. There are multiple ways of realizing $p'_2 < p_2$. In one example, only the relay transmits packet in the second round, as the the access link is assumed to have a better channel than the macro-access link. In another example, both the relay and the base station transmits the same signals, so that the mobile terminal receiver can get benefits from increased receive power.

III. COORDINATED MULTI-POINT TRANSMISSIONS

In this section, two important classes of CoMP transmission schemes are introduced and discussed.

A. Joint Processing

In the class of joint processing/transmission, multiple base stations (or transmission points or cells) jointly transmit signals to a single mobile terminal improve the received signal quality or actively cancel interference for other mobile terminals, or both. In this case, data intended for a particular mobile terminal is shared among different cells (cell 1 and cell 2) and is jointly processed at these cells. As a result of this joint processing, received signals at the intended mobile terminal will be coherently or non-coherently added up together.

We assume that mobile user 1 (M1) is receiving signals from the three cells: Cell 1, Cell 2, and Cell 3 (denoted as C1, C2 and C3). Assume H_{i1} is the channel gain from Ci to M1, the received signal Y_1 at M1 can be expressed as

$$Y_1 = H_{11}W_1X_1 + H_{21}W_2X_2 + H_{31}W_3X_3 + Z_1,$$

where X_i is the signal transmitted at C_i , W_i is the precoding matrix at C_i , and Z_1 is the additive white Gaussian noise at the receiver. If each cell is serving to his/her own mobile terminals, the signals will interference with each other, then the SINR for M1 can be expressed as

$$SINR_1 = \frac{||H_{11}W_1||^2 P_1}{||H_{21}W_2||^2 P_2 + ||H_{31}W_3||^2 P_3 + N}$$

where P_i is the transmitted power of X_i at C_i , and N is the noise power. Consider a CoMP joint processing system where C1, C2 and C3 form a CoMP cluster, M1 is then being simultaneously served by all the three cells belonging to the CoMP cluster. Under this assumption, we have $X_1 = X_2 = X_3 = X$. Accordingly, the received signal at M1 can be expressed as

$$Y_1 = H_{11}W_1X + H_{21}W_2X + H_{31}W_3X + Z_1,$$

Therefore, the SINR of M1 can be computed as

$$\operatorname{SINR}_{1}' = \frac{||H_{11}W_{1}\sqrt{P_{1}} + H_{21}W_{2}\sqrt{P_{2}} + H_{31}W_{3}\sqrt{P_{3}}||^{2}}{N}$$

It is clear that $SINR_1$ is always upper-bounded by $SINR'_1$, and CoMP joint processing will always bring a SINR gain compared to single-cell operation. However, this gain is not free. Note that $SINR_1$ is obtained under the assumption that each cell is serving his/her own mobile user while $SINR'_1$ is obtained under the assumption that three cells are serving one mobile user. Therefore, mobile terminals under CoMP joint processing are occupying more system resources than the single-cell mobile terminals. This is actually one of the biggest hidden costs of CoMP joint processing. Taking this hidden cost in to account, assuming symmetric channel conditions, the total throughputs of single-cell operation and CoMP joint processing are

$$3 \log_2 (1 + \text{SINR}_1)$$
 and $\log_2 (1 + \text{SINR}'_1)$

respectively. Therefore, it does not worth to perform CoMP joint processing for cell-center mobile terminals where the value of $SINR_1$ is high. Under this situation, the cost of system resource is high while SINR improvement is marginal.

The above analysis is based on the fact only one mobile user is served by the CoMP cluster which is called CoMP singleuser (SU) MIMO mode. A more involved operation mode is the CoMP multi-user (MU) MIMO mode where multiple mobile terminals are joint served by the CoMP cluster.

B. Coordinated Beamforming/Scheduling

In the class of coordinated scheduling and/or beam-forming, data to mobile terminal is instantaneously transmitted from one of the transmission points (cells) while the scheduling decisions are coordinated to control the interference generated in a set of coordinated cells. In other words, the data intended for a particular mobile terminal, say M1, is transmitted only by C1; however, C2 will choose to serve its mobile terminals in such a way that it will create little interference mitigation" in the signal processing society and some methods to mitigate interference through different signal spaces can be found [11], [12].

Assume two mobile terminals, M1 and M2, are close to each other and are served by C1 and C2 respectively. The received signals, Y_1 and Y_2 , of M1 and M2 can be written as

$$Y_1 = H_{11}W_1X_1 + H_{21}W_2X_2 + Z_1$$

$$Y_2 = H_{12}W_1X_1 + H_{22}W_2X_2 + Z_2.$$

Accordingly, the received SINR for M1 and M2 can be expressed as

$$SINR_1 = \frac{||H_{11}W_1||^2 P_1}{||H_{21}W_2||^2 P_2 + N}$$

$$SINR_2 = \frac{||H_{22}W_2||^2 P_2}{||H_{12}W_1||^2 P_1 + N}.$$

In single-cell operation, the precoding vector W_i of MS *i* is chose such that the received signal strength from the serving cell are maximized:

When M1 and M2 are close, it is likely that both the pairs $\{H_{11}, H_{12}\}$ and $\{H_{21}, H_{22}\}$ are correlated. Therefore, W'_1 applied in C1 is actually causing a large inter-cell interference to the received signal of M2 and vice versa. In coordinated beamforming/scheduling, the precoding vectors are joint optimized such that the SINRs at the mobile terminals are improved. That is, the CoMP cluster joint choose the precoding vectors and scheduling decisions taking into account the inter-cell interference.

C. Joint Processing vs. Coordinated Beamforming/Scheduling

It is expected that CoMP joint processing will bring more significant system improvement at a higher implementation cost. For example, in CoMP joint processing, the data together with channel related information for different mobile users needs to be exchanged among the cells within CoMP cluster. Differently from relaying technology, this data exchange can be done in wired backhaul; however, this still cause additional latency and impose stringent requirements for backhaul technologies. On the other hand, in coordinated beamforming/scheduling, unlike joint processing, data for an intended mobile user is only transmitted from its serving cell. This way, only channel state information and scheduling decisions are needed to be exchanged among the cells. This reduces the system complexity and backhaul traffic. Furthermore, joint processing is more sensitive to the channel feedback errors as opposed to the coordinated beamforming/scheduling, and it is difficult to ensure that the signals from different cells are constructively add at the receiver. Due to these reasons, current 3GPP community focuses more on the CoMP coordinated beamforming/scheduling for LTE-A systems. However, joint processing can be still promising if some of the practical issues can be resolved.

IV. CONCLUSION

In this paper, cooperative communication technologies being considered in LTE-A have been introduced. As solutions for coverage limitation and inter-cell interference, relaying and coordinated multi-point (CoMP) transmission have respectively been discussed, and their performances and design challenges have been investigated.

REFERENCES

- 3GPP TR 36814, Evolved Universal Terrestrial Radio Access (E-UTRA); Further advancements for E-UTRA Physical layer aspects Physical channels and modulation, ver 1.0.0, Feb. 2009.
- [2] 3GPP TS 36211, Evolved Universal Terrestrial Radio Access (E-UTRA); Physical channels and modulation, ver 8.7.0, Jun. 2009.
- [3] T. M. Cover and J. A Thomas, *Elements of Information Theory*, Wiley, 1991.
- [4] D. Tse and P. Viswanath, Fundamentals of Wireless Communications, Cambridge University Press, 2005.
- [5] H. Zhang, N. B. Mehta, A. F. Molisch, J. Zhang, and H. Dai, "Asynchronous Interference Mitigation in Cooperative Base Station Systems," *IEEE Tr. Wireless Comm.*, Vol. 7, Issue 1, pp. 155-165, January 2008.
- [6] P. Gupta and P. R. Kumar, "The Capacity of Wireless Networks," *IEEE Trans. Inform. Theory*, Vol. 46, No. 3, pp. 388–404, Mar. 2000.
- [7] A. Sendonaris, E. Erkip and B. Aazhang, "User Cooperation Diversity -Part I: System Description," *IEEE Trans. Comm.*, Vol. 51, No. 11, Nov. 2003.
- [8] Y. H. Nam, K. Azarian, H. El Gamal and P. Schniter "Cooperation through ARQ," Sig. Proc. Adv. Wireless Comm. (SPAWC), Jun. 2005.
- [9] J. N. Laneman, D. N. C. Tse and G. W. Wornell, "Cooperative Diversity in Wireless Networks: Efficient Protocols and Outage Behavior," *IEEE Trans. Inform. Theory*, vol. 50, no. 12, pp. 3062-3080, Dec. 2004.
- [10] K. Azarian, H. El Gamal and P. Schniter, "On the Achievable Diversity-Multiplexing Tradeoff in Half-Duplex Cooperative Channels," *IEEE Trans. Inform. Theory*, vol. 51, no. 12, pp. 4152-4172, Dec. 2005.
- [11] J. Kotecha and J. Mundarath, "Non-Collaborative Zero-Forcing Beamforming in the Presence of Co-Channel Interference and Spatially Correlated Channels," in Proc. of *IEEE VTC*, pp. 591-595, Fall 2007.
- [12] L. Liu, J. Zhang, J. Yu and J. Lee, "Inter-Cell Interference Coordination Through Limited Feedback," Special Issue on Multicell Cooperation and MIMO Technologies for Broadcasting and Broadband Communications, *International Journal of Digital Multimedia Broadcasting*, 2009.